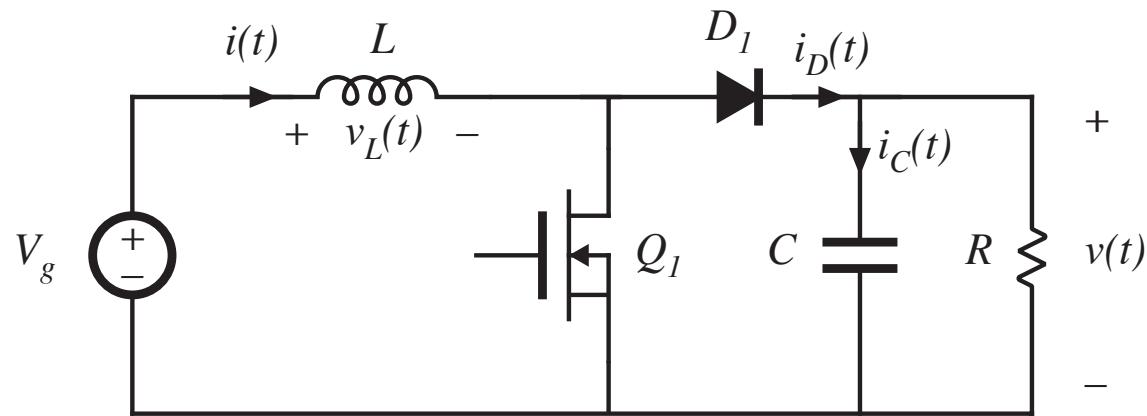


5.3. Boost converter example



Mode boundary:

$$\begin{aligned} I > \Delta i_L &\text{ for CCM} \\ I < \Delta i_L &\text{ for DCM} \end{aligned}$$

Previous CCM soln:

$$I = \frac{V_g}{D'^2 R} \quad \Delta i_L = \frac{V_g}{2L} DT_s$$

Mode boundary

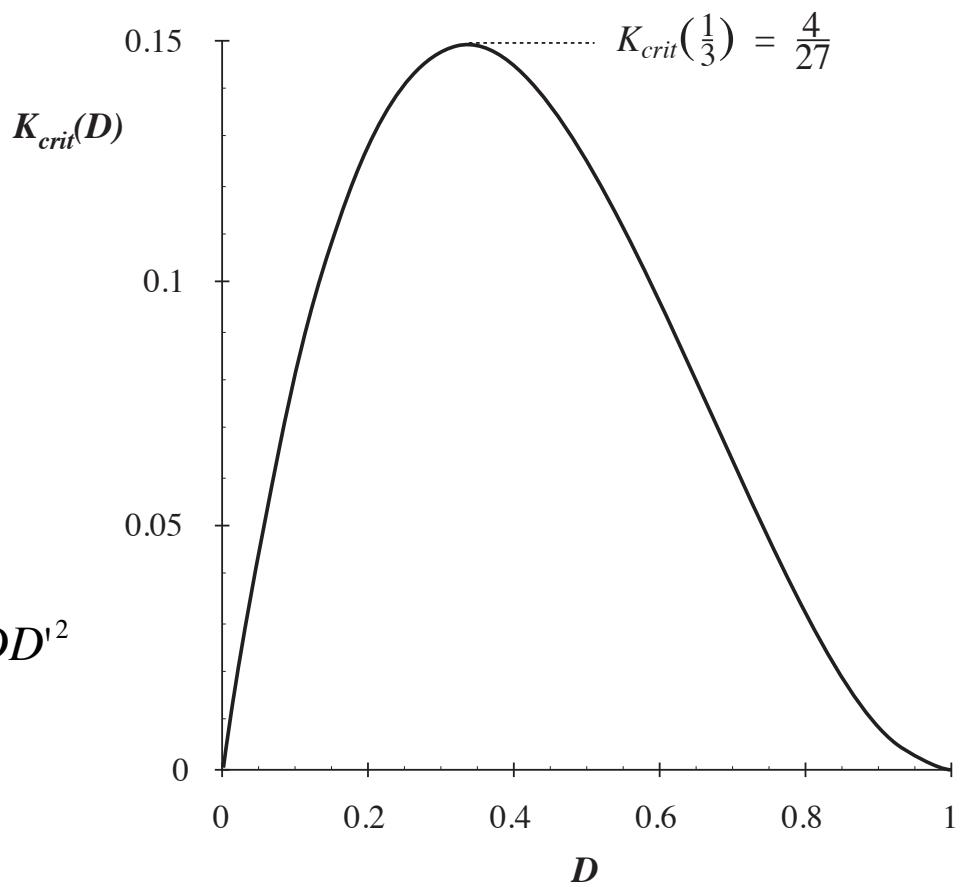
$$\frac{V_g}{D'^2 R} > \frac{DT_s V_g}{2L} \quad \text{for CCM}$$

$$\frac{2L}{RT_s} > DD'^2 \quad \text{for CCM}$$

$K > K_{crit}(D)$ for CCM

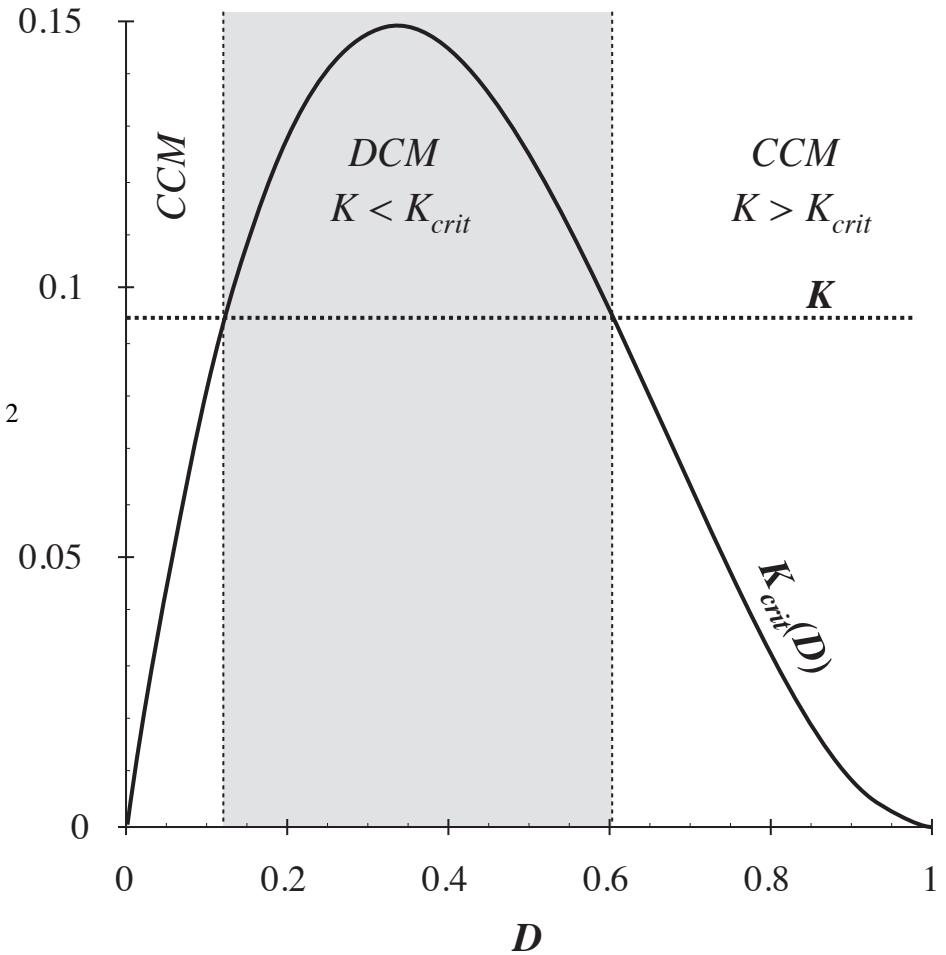
$K < K_{crit}(D)$ for DCM

where $K = \frac{2L}{RT_s}$ and $K_{crit}(D) = DD'^2$

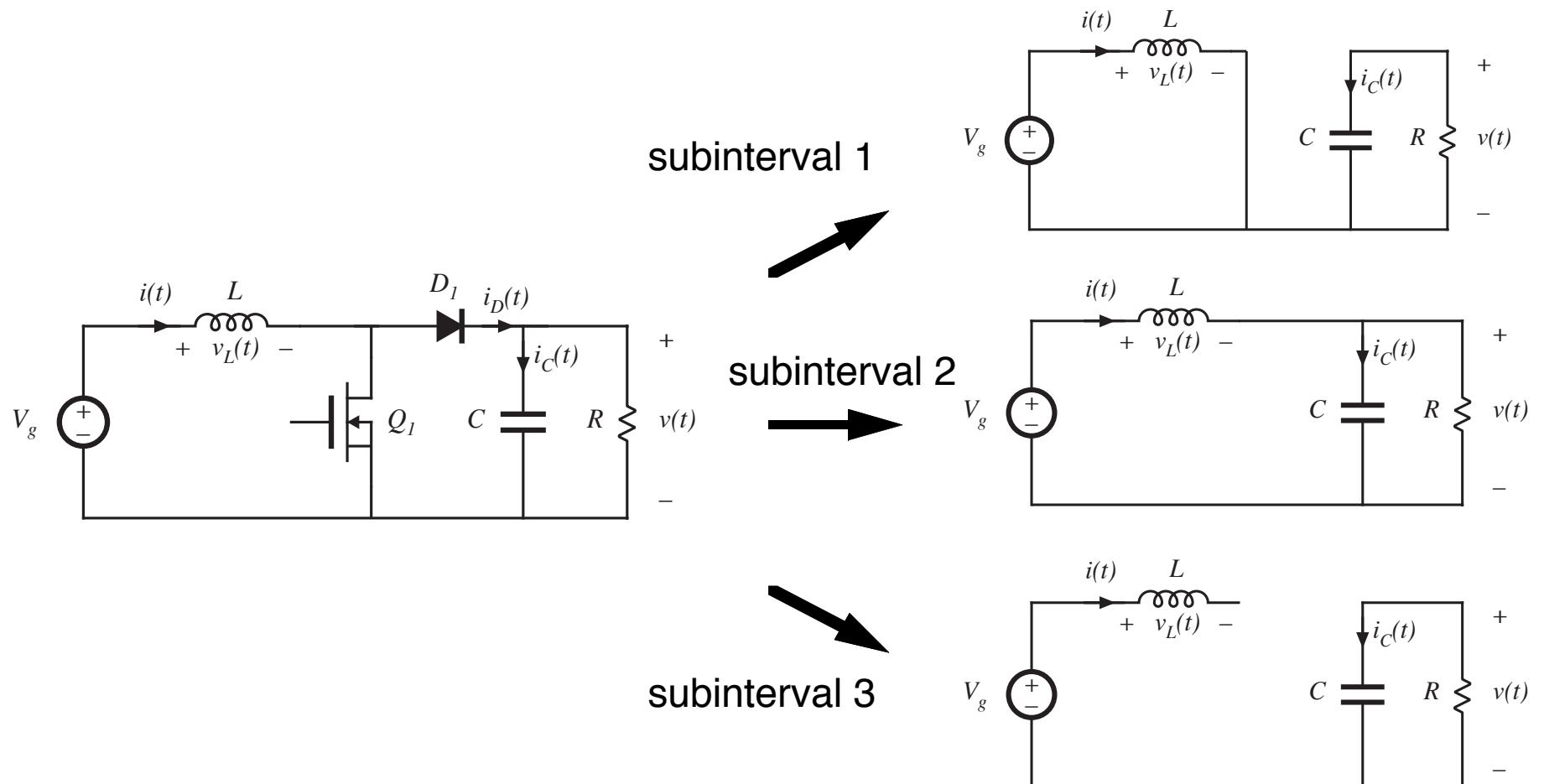


Mode boundary

$K > K_{crit}(D)$ for CCM
 $K < K_{crit}(D)$ for DCM
where $K = \frac{2L}{RT_s}$ and $K_{crit}(D) = DD'^2$



Conversion ratio: DCM boost

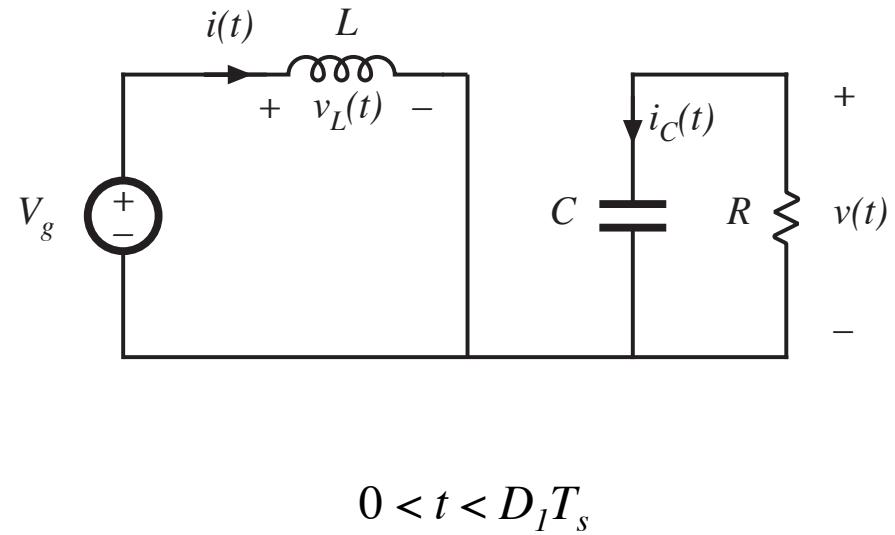


Subinterval 1

$$v_L(t) = V_g$$
$$i_C(t) = -v(t) / R$$

Small ripple approximation
for $v(t)$ (but not for $i(t)$!):

$$v_L(t) \approx V_g$$
$$i_C(t) \approx -V / R$$



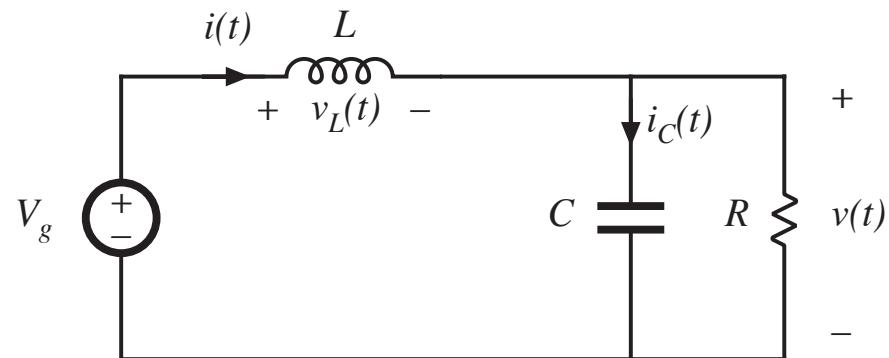
$$0 < t < D_1 T_s$$

Subinterval 2

$$v_L(t) = V_g - v(t)$$
$$i_C(t) = i(t) - v(t) / R$$

Small ripple approximation
for $v(t)$ but not for $i(t)$:

$$v_L(t) \approx V_g - V$$
$$i_C(t) \approx i(t) - V / R$$



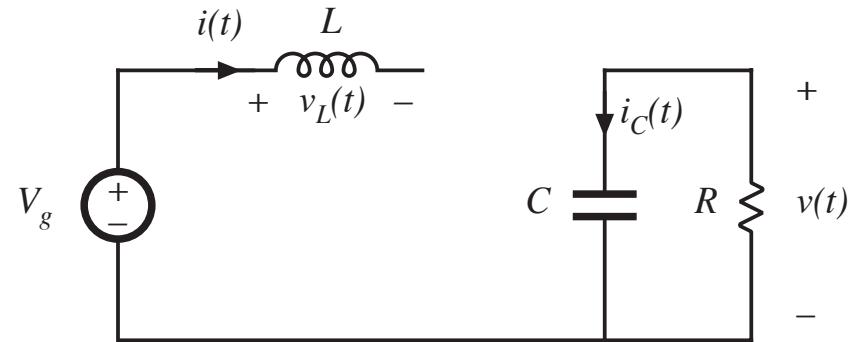
$$D_1 T_s < t < (D_1 + D_2) T_s$$

Subinterval 3

$$v_L = 0, \quad i = 0 \\ i_C(t) = -v(t) / R$$

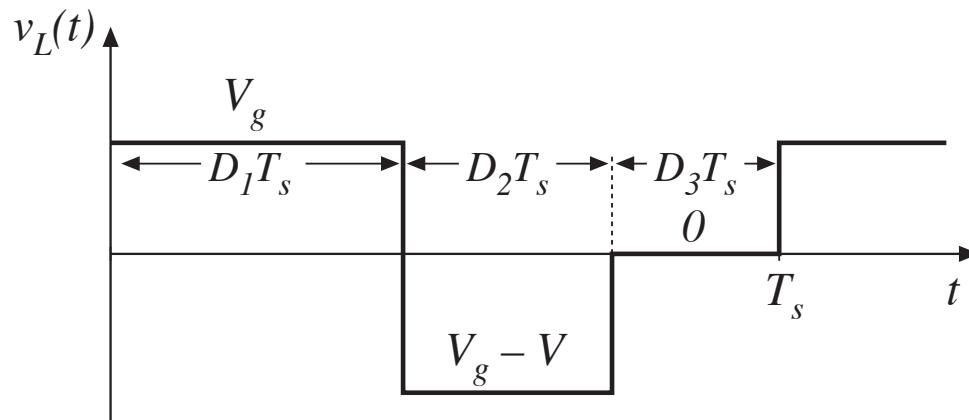
Small ripple approximation:

$$v_L(t) = 0 \\ i_C(t) = -V / R$$



$$(D_1 + D_2)T_s < t < T_s$$

Inductor volt-second balance



Volt-second balance:

$$D_1 V_g + D_2(V_g - V) + D_3(0) = 0$$

Solve for V :

$$V = \frac{D_1 + D_2}{D_2} V_g$$

note that D_2 is unknown

Capacitor charge balance

node equation:

$$i_D(t) = i_C(t) + v(t) / R$$

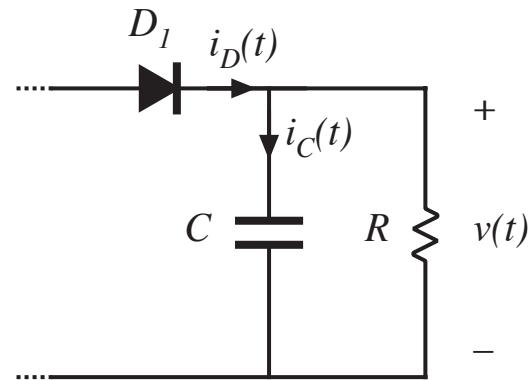
capacitor charge balance:

$$\langle i_C \rangle = 0$$

hence

$$\langle i_D \rangle = V / R$$

must compute dc component of diode current and equate to load current
(for this boost converter example)



Inductor and diode current waveforms

peak current:

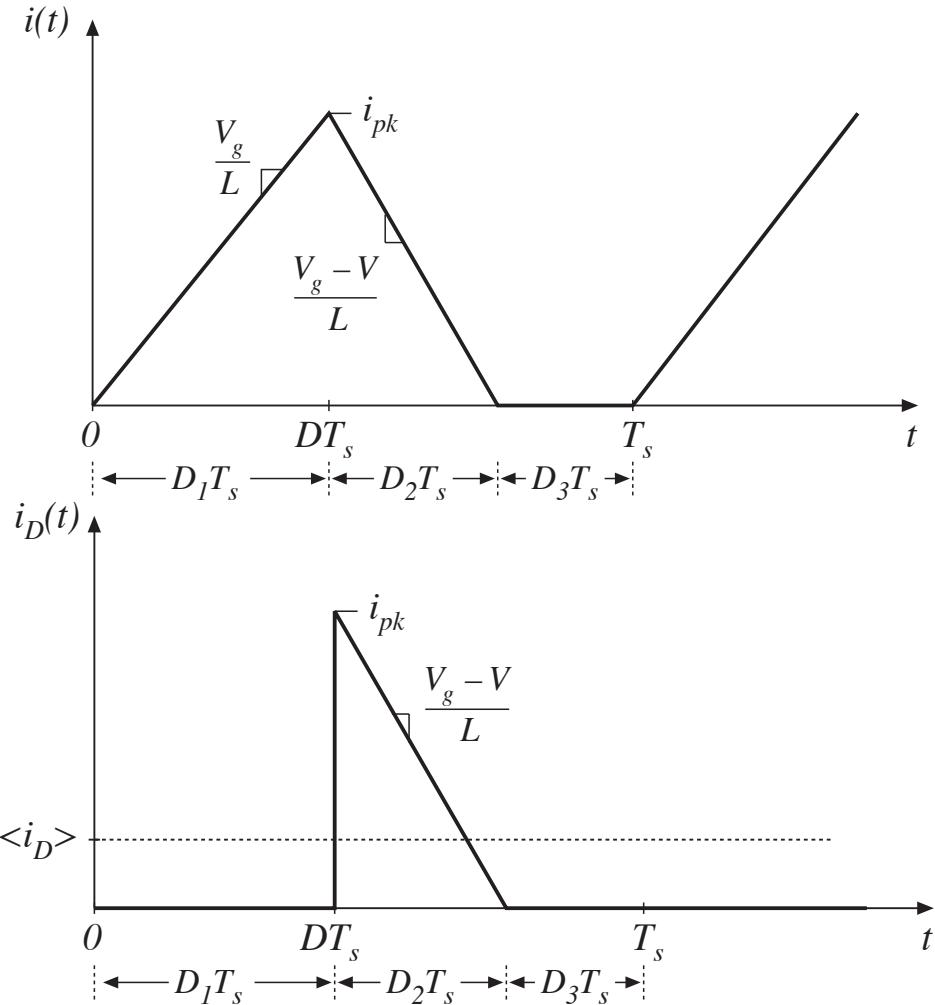
$$i_{pk} = \frac{V_g}{L} D_1 T_s$$

average diode current:

$$\langle i_D \rangle = \frac{1}{T_s} \int_0^{T_s} i_D(t) dt$$

triangle area formula:

$$\int_0^{T_s} i_D(t) dt = \frac{1}{2} i_{pk} D_2 T_s$$



Equate diode current to load current

average diode current:

$$\langle i_D \rangle = \frac{1}{T_s} \left(\frac{1}{2} i_{pk} D_2 T_s \right) = \frac{V_g D_1 D_2 T_s}{2L}$$

equate to dc load current:

$$\frac{V_g D_1 D_2 T_s}{2L} = \frac{V}{R}$$

Solution for V

Two equations and two unknowns (V and D_2):

$$V = \frac{D_1 + D_2}{D_2} V_g \quad (\text{from inductor volt-second balance})$$

$$\frac{V_g D_1 D_2 T_s}{2L} = \frac{V}{R} \quad (\text{from capacitor charge balance})$$

Eliminate D_2 , solve for V . From volt-sec balance eqn:

$$D_2 = D_1 \frac{V_g}{V - V_g}$$

Substitute into charge balance eqn, rearrange terms:

$$V^2 - VV_g - \frac{V_g^2 D_1^2}{K} = 0$$

Solution for V

$$V^2 - VV_g - \frac{V_g^2 D_1^2}{K} = 0$$

Use quadratic formula:

$$\frac{V}{V_g} = \frac{1 \pm \sqrt{1 + 4D_1^2 / K}}{2}$$

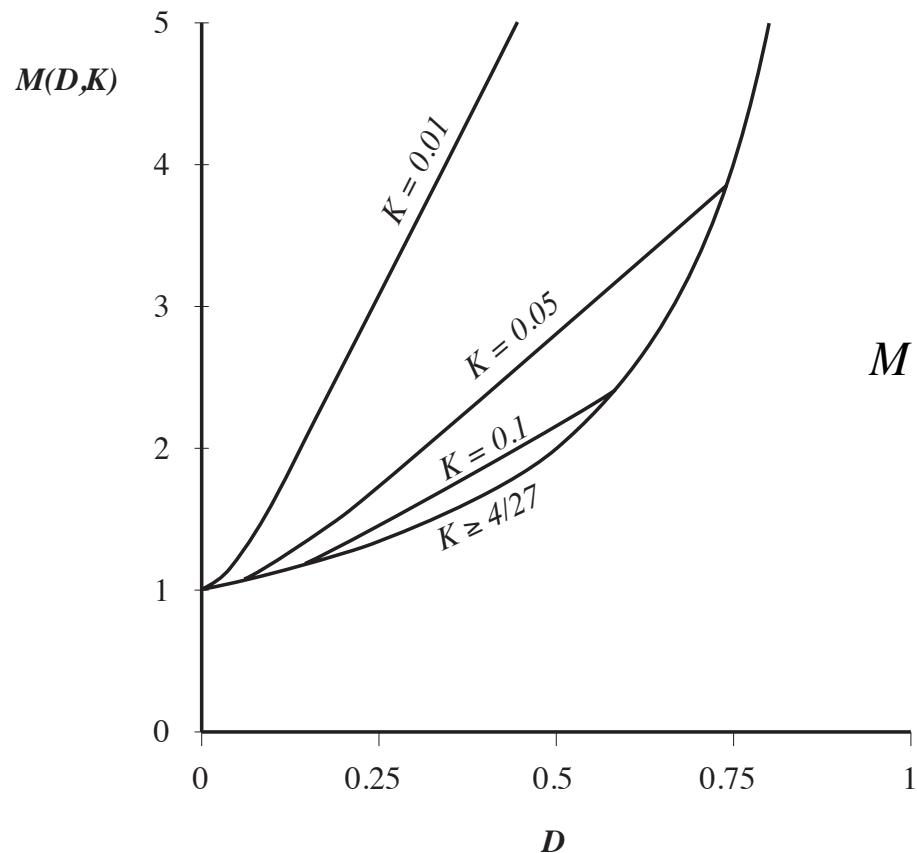
Note that one root leads to positive V , while other leads to negative V . Select positive root:

$$\frac{V}{V_g} = M(D_1, K) = \frac{1 + \sqrt{1 + 4D_1^2 / K}}{2}$$

where $K = 2L / RT_s$
valid for $K < K_{crit}(D)$

Transistor duty cycle D = interval 1 duty cycle D_1

Boost converter characteristics



$$M = \begin{cases} \frac{1}{1-D} & \text{for } K > K_{crit} \\ \frac{1 + \sqrt{1 + 4D^2 / K}}{2} & \text{for } K < K_{crit} \end{cases}$$

Approximate M in DCM:

$$M \approx \frac{1}{2} + \frac{D}{\sqrt{K}}$$

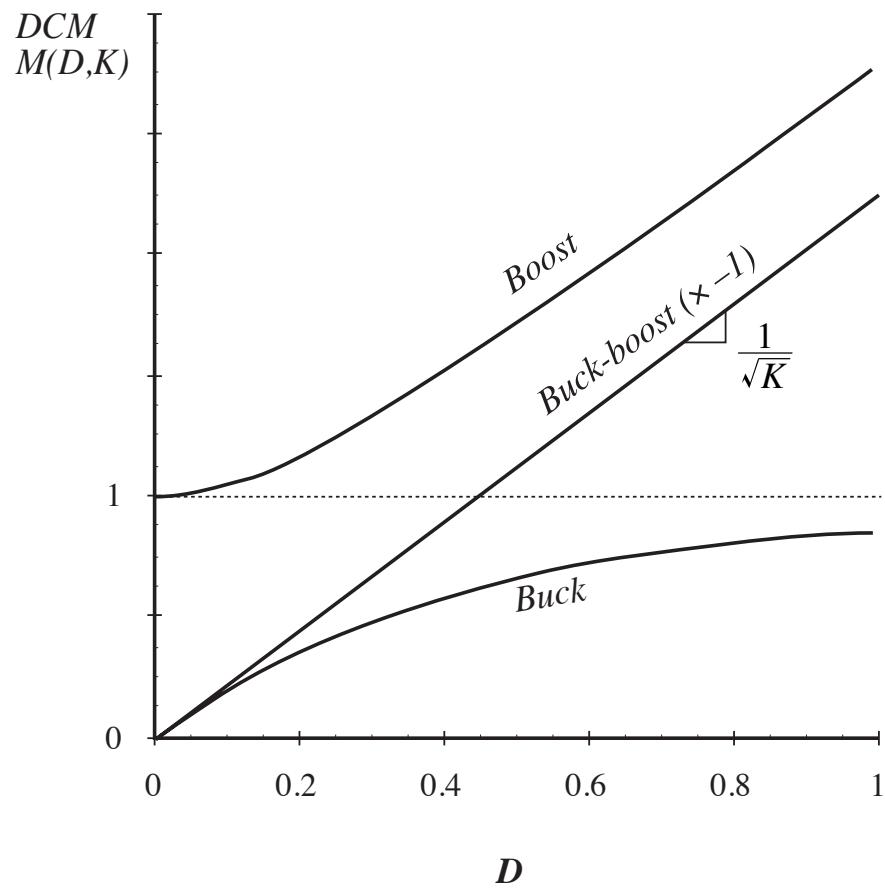
Summary of DCM characteristics

Table 5.2. Summary of CCM-DCM characteristics for the buck, boost, and buck-boost converters

<i>Converter</i>	$K_{crit}(D)$	<i>DCM M(D,K)</i>	<i>DCM D₂(D,K)</i>	<i>CCM M(D)</i>
Buck	$(1 - D)$	$\frac{2}{1 + \sqrt{1 + 4K / D^2}}$	$\frac{K}{D} M(D,K)$	D
Boost	$D (1 - D)^2$	$\frac{1 + \sqrt{1 + 4D^2 / K}}{2}$	$\frac{K}{D} M(D,K)$	$\frac{1}{1 - D}$
Buck-boost	$(1 - D)^2$	$-\frac{D}{\sqrt{K}}$	\sqrt{K}	$-\frac{D}{1 - D}$

with $K = 2L / RT_s$. DCM occurs for $K < K_{crit}$.

Summary of DCM characteristics



- DCM buck and boost characteristics are asymptotic to $M = 1$ and to the DCM buck-boost characteristic
- DCM buck-boost characteristic is linear
- CCM and DCM characteristics intersect at mode boundary. Actual M follows characteristic having larger magnitude
- DCM boost characteristic is nearly linear