## 9.4.2. The relation between phase margin and closed-loop damping factor

How much phase margin is required?

A small positive phase margin leads to a stable closed-loop system having complex poles near the crossover frequency with high *Q*. The transient response exhibits overshoot and ringing.

Increasing the phase margin reduces the *Q*. Obtaining real poles, with no overshoot and ringing, requires a large phase margin.

The relation between phase margin and closed-loop *Q* is quantified in this section.

#### A simple second-order system



# Closed-loop response

If

$$
T(s) = \frac{1}{\left(\frac{s}{\omega_0}\right)\left(1 + \frac{s}{\omega_2}\right)}
$$

Then

$$
\frac{T(s)}{1+T(s)} = \frac{1}{1+\frac{1}{T(s)}} = \frac{1}{1+\frac{s}{\omega_0}+\frac{s^2}{\omega_0\omega_2}}
$$

or,

$$
\frac{T(s)}{1+T(s)} = \frac{1}{1+\frac{s}{Q\omega_c}+\left(\frac{s}{\omega_c}\right)^2}
$$

where

$$
\omega_c = \sqrt{\omega_0 \omega_2} = 2\pi f_c \qquad \qquad Q = \frac{\omega_0}{\omega_c} = \sqrt{\frac{\omega_0}{\omega_2}}
$$

*Fundamentals of Power Electronics Chapter 9: Controller design* 30

## Low-*Q* case



# High-*Q* case





$$
Q
$$
 vs.  $\varphi_m$ 

Solve for exact crossover frequency, evaluate phase margin, express as function of  $\operatorname{\phi_{m}}$ . Result is:

$$
Q = \frac{\sqrt{\cos \varphi_m}}{\sin \varphi_m}
$$

$$
\varphi_m = \tan^{-1} \sqrt{\frac{1 + \sqrt{1 + 4Q^4}}{2Q^4}}
$$

*Fundamentals of Power Electronics* 33 3 3 *Chapter 9: Controller design* 







### 9.4.3. Transient response vs. damping factor

Unit-step response of second-order system *T*(*s*)/(1+*T*(*s*))

$$
\hat{v}(t) = 1 + \frac{2Q e^{-\omega_c t/2Q}}{\sqrt{4Q^2 - 1}} \sin \left[ \frac{\sqrt{4Q^2 - 1}}{2Q} \omega_c t + \tan^{-1} \left( \sqrt{4Q^2 - 1} \right) \right] \qquad Q > 0.5
$$

$$
\hat{v}(t) = 1 - \frac{\omega_2}{\omega_2 - \omega_1} e^{-\omega_1 t} - \frac{\omega_1}{\omega_1 - \omega_2} e^{-\omega_2 t}
$$
\n
$$
\omega_1, \omega_2 = \frac{\omega_c}{2Q} \left( 1 \pm \sqrt{1 - 4Q^2} \right)
$$
\n
$$
\omega_2 = \frac{\omega_c}{2Q} \left( 1 - \sqrt{1 - 4Q^2} \right)
$$
\n
$$
\omega_1 = \frac{\omega_2}{2Q} \left( 1 - \sqrt{1 - 4Q^2} \right)
$$

For  $Q > 0.5$ , the peak value is

peak 
$$
\hat{v}(t) = 1 + e^{-\pi/\sqrt{4Q^2 - 1}}
$$

*Fundamentals of Power Electronics* 35 35 *Chapter 9: Controller design* 

### Transient response vs. damping factor



*Fundamentals of Power Electronics* 36 36 *Chapter 9: Controller design* 

